

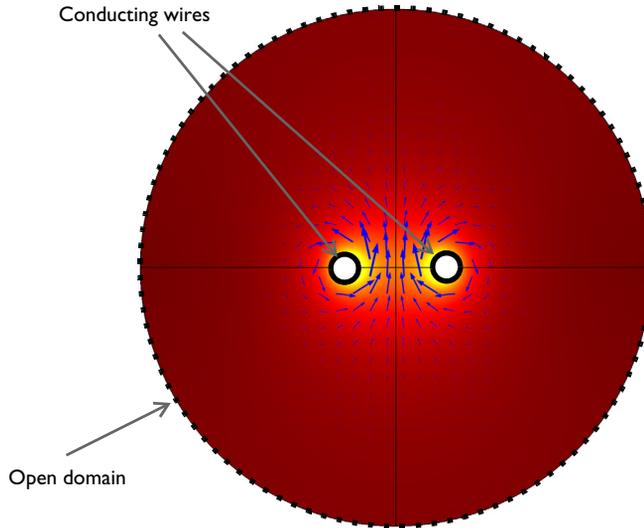


# Finding the Impedance of a Parallel-Wire Transmission Line

## Introduction

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A parallel-wire transmission line is composed of two conducting wires in a dielectric such as air. The fields around such a transmission line are not directly confined by the conductors but extend to infinity, although they drop off rapidly away from the wires. This example demonstrates how to compute the fields and impedance of such an unshielded transmission line and compares the results to the analytic solution.



*Figure 1: A parallel-wire transmission line. The electric field strength is plotted in color while arrows show the magnetic field.*

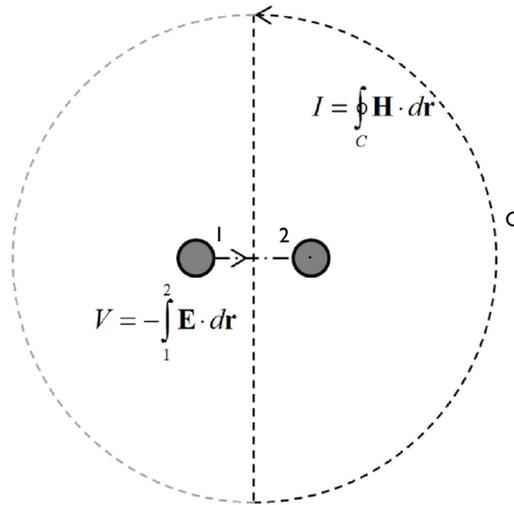
## Model Definition

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Because a parallel-wire transmission line operates in TEM mode—with the electric and magnetic fields normal to the direction of propagation along the cable—modeling a 2D cross section suffices to compute the fields and the impedance. For this example, assume perfect conductors and a lossless air region. The wires, of radius is 1 mm, are separated by a center-to-center distance of 8 mm.

Because the structure is open, the fields extend infinitely far away from the wires. However, they drop off quickly in magnitude. This raises the question about what boundary condition to use on the air domain's outer boundary. The surrounding dielectric medium

can be thought of as a perfect insulator, as opposed to the wires, which are modeled as perfect conductors. Thus, the model uses a perfect magnetic conductor (PMC) boundary condition, because this condition is, in a sense, the opposite of the perfect electric conductor (PEC) boundary condition. However, it must be placed some distance away from the wires or else it would artificially confine the fields. In this example, the air domain's radius is chosen to be five times the distance from its center to the center of each wire. Increasing this radius would give a more accurate solution at the cost of a more memory-intensive computation.



*Figure 2: The impedance of a parallel-wire transmission line can be found from the voltage,  $V$ , and current,  $I$ , which are computed via line integrals as shown.*

The characteristic impedance,  $Z_0 = V/I$ , of a transmission line relates the voltage to the current. Although the model does not involve computing the potential field, the voltage of the TEM waveguide can be evaluated as a line integral of the electric field between the conductors:

$$V = V_2 - V_1 = -\int_1^2 \mathbf{E} \cdot d\mathbf{r} \quad (1)$$

Similarly, the current is obtained as a line integral of the magnetic field along the boundary of either conductor, or any closed contour,  $C$ , bisecting the space between the conductors:

$$I = \oint_C \mathbf{H} \cdot d\mathbf{r}$$

The voltage and current in the direction out of the plane are positive for integration paths oriented as in [Figure 2](#).

The value of  $Z_0$  obtained in this way, should be compared with the analytic result

$$Z_{0, \text{analytic}} = \frac{1}{\pi} \sqrt{\frac{\mu_0}{\epsilon_0 \epsilon_r}} \operatorname{acosh}\left(\frac{r_d}{r_a}\right) = 247 \Omega$$

Here  $r_a$  is the wire radius and  $r_d$  is the center-to-center distance between the wires.

## Results and Discussion

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Figure 3 is a combined plot of the electric field magnitude and the magnetic field visualized as an arrow plot.

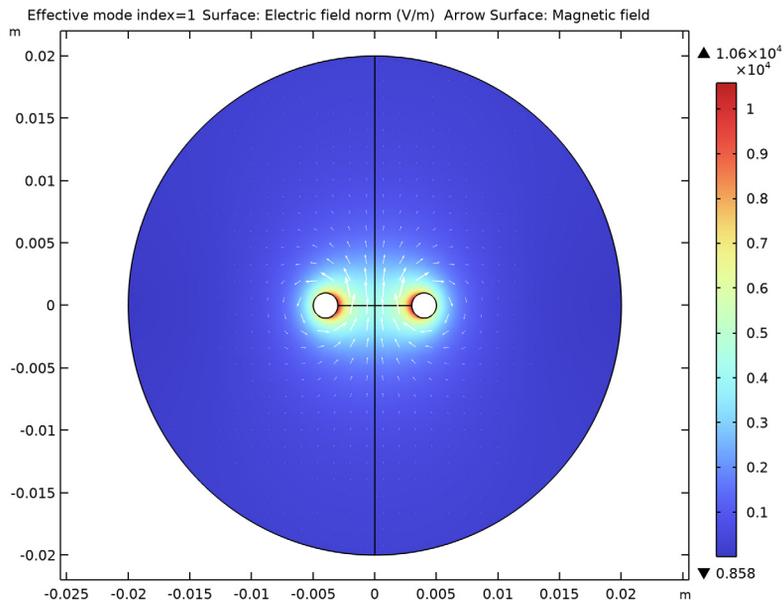


Figure 3: Electric field magnitude (surface) and magnetic field (arrows) around the two parallel wires.

The impedance computed with the default mesh is  $Z_0 = 255.8 \Omega$ . As the radius of the dielectric domain is increased, the numerical solution will approach the analytic value of  $247.4 \Omega$ .

### Notes About the COMSOL Implementation

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Solve this example using a Mode Analysis study and the default frequency,  $f = 1$  GHz.

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**Application Library path:** RF\_Module/Verification\_Examples/  
parallel\_wires\_impedance

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## Modeling Instructions

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From the **File** menu, choose **New**.

### NEW

In the **New** window, click  **Model Wizard**.

### MODEL WIZARD

- 1 In the **Model Wizard** window, click  **2D**.
- 2 In the **Select Physics** tree, select **Radio Frequency>Electromagnetic Waves, Frequency Domain (emw)**.
- 3 Click **Add**.
- 4 Click  **Study**.
- 5 In the **Select Study** tree, select **Preset Studies for Selected Physics Interfaces>Mode Analysis**.
- 6 Click  **Done**.

### GLOBAL DEFINITIONS

#### Parameters 1

- 1 In the **Model Builder** window, under **Global Definitions** click **Parameters 1**.
- 2 In the **Settings** window for **Parameters**, locate the **Parameters** section.
- 3 In the table, enter the following settings:

Name	Expression	Value	Description
r_a	1[mm]	0.001 m	Wire radius
r_d	4[mm]	0.004 m	Center-to-center distance between wires
r_air	5*r_d	0.02 m	Air-domain radius
Z0_analytic	$(Z0\_const/pi)*\log(r\_d/r\_a+sqrt((r\_d/r\_a)^2-1))$	247.44 $\Omega$	Characteristic impedance, analytic

Here,  $Z0\_const$  is a predefined COMSOL constant for the characteristic impedance of vacuum,  $Z0 = \sqrt{\mu_0/\epsilon_0}$ . From the Value column you can read off the value  $Z0$ , analytic = 247  $\Omega$ . Note also that the logarithm in the definition for  $Z0\_analytic$  is an equivalent way of writing  $\text{acosh}(r\_d/r\_a)$ .

## GEOMETRY 1

First, create a circle for the air domain.

### Circle 1 (c1)

- 1 In the **Geometry** toolbar, click  **Circle**.
- 2 In the **Settings** window for **Circle**, locate the **Size and Shape** section.
- 3 In the **Radius** text field, type `r_air`.
- 4 Click  **Build Selected**.

Add a circle for one wire.

### Circle 2 (c2)

- 1 In the **Geometry** toolbar, click  **Circle**.
- 2 In the **Settings** window for **Circle**, locate the **Size and Shape** section.
- 3 In the **Radius** text field, type `r_a`.
- 4 Locate the **Position** section. In the **x** text field, type `r_d`.
- 5 Click  **Build Selected**.

Then, generate the other wire by mirroring the above one.

### Mirror 1 (mir1)

- 1 In the **Geometry** toolbar, click  **Transforms** and choose **Mirror**.
- 2 Select the object **c2** only.
- 3 In the **Settings** window for **Mirror**, locate the **Input** section.
- 4 Select the **Keep input objects** check box.
- 5 Click  **Build Selected**.

### Difference 1 (dif1)

- 1 In the **Geometry** toolbar, click  **Booleans and Partitions** and choose **Difference**.
- 2 Select the object **c1** only.
- 3 In the **Settings** window for **Difference**, locate the **Difference** section.
- 4 Find the **Objects to subtract** subsection. Click to select the  **Activate Selection** toggle button.
- 5 Select the objects **c2** and **mir1** only.
- 6 Click  **Build Selected**.

Create a line for computing the voltage as a line integral of the electric field.

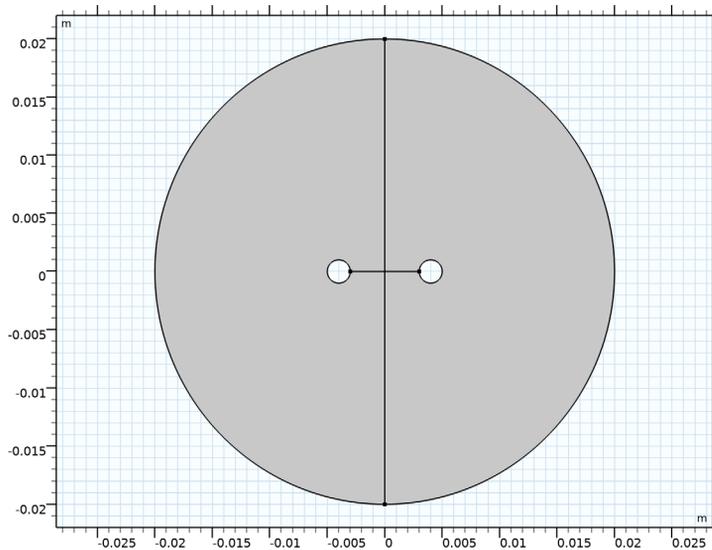
### Line Segment 1 (ls1)

- 1 In the **Geometry** toolbar, click  **More Primitives** and choose **Line Segment**.
- 2 In the **Settings** window for **Line Segment**, locate the **Starting Point** section.
- 3 From the **Specify** list, choose **Coordinates**.
- 4 Locate the **Endpoint** section. From the **Specify** list, choose **Coordinates**.
- 5 Locate the **Starting Point** section. In the **x** text field, type  $-r_d+r_a$ .
- 6 Locate the **Endpoint** section. In the **x** text field, type  $r_d-r_a$ .

Add a line, a part of the closed contour for computing the current as a line integral of the magnetic field.

### Line Segment 2 (ls2)

- 1 In the **Geometry** toolbar, click  **More Primitives** and choose **Line Segment**.
- 2 In the **Settings** window for **Line Segment**, locate the **Starting Point** section.
- 3 From the **Specify** list, choose **Coordinates**.
- 4 Locate the **Endpoint** section. From the **Specify** list, choose **Coordinates**.
- 5 Locate the **Starting Point** section. In the **y** text field, type  $-r_{air}$ .
- 6 Locate the **Endpoint** section. In the **y** text field, type  $r_{air}$ .
- 7 Click  **Build Selected**.



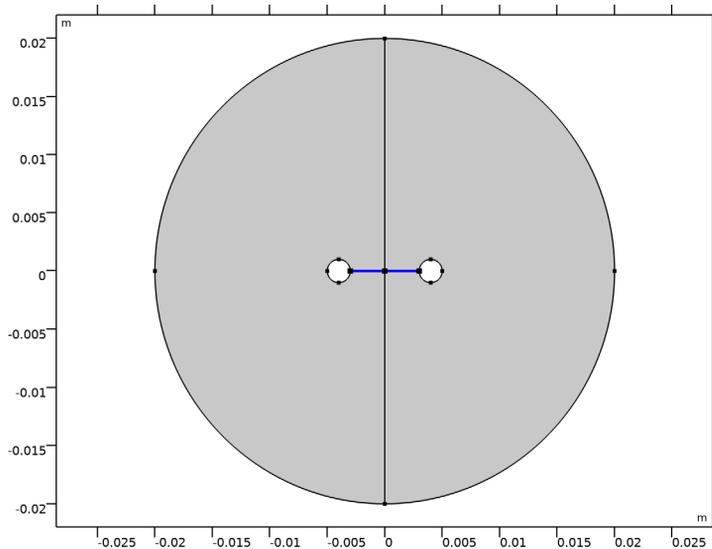
The model layout describes two parallel wires in the air.

## DEFINITIONS

Add a variable for the characteristic impedance computed as the voltage between the wires divided by the current through the wires. Define two nonlocal integration couplings for computing the voltage and the current.

### *Integration 1 (intop1)*

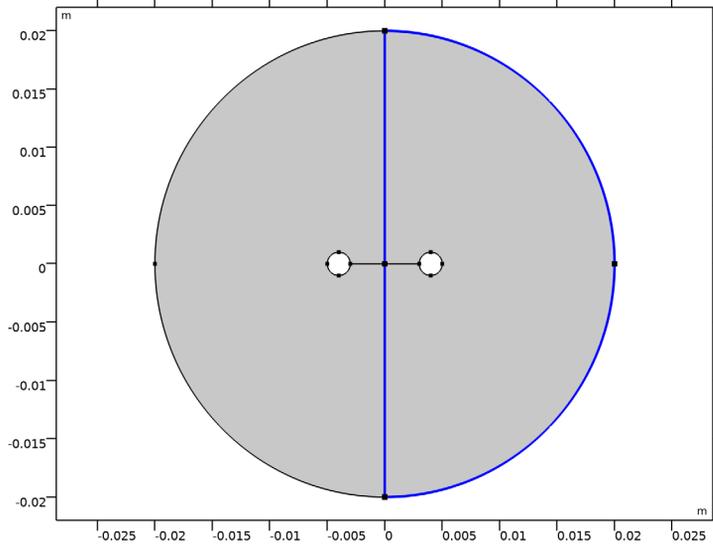
- 1 In the **Definitions** toolbar, click  **Nonlocal Couplings** and choose **Integration**.
- 2 In the **Settings** window for **Integration**, type `int_E` in the **Operator name** text field.
- 3 Locate the **Source Selection** section. From the **Geometric entity level** list, choose **Boundary**.
- 4 Select Boundaries 1 and 4 only.



### *Integration 2 (intop2)*

- 1 In the **Definitions** toolbar, click  **Nonlocal Couplings** and choose **Integration**.
- 2 In the **Settings** window for **Integration**, type `int_H` in the **Operator name** text field.
- 3 Locate the **Source Selection** section. From the **Geometric entity level** list, choose **Boundary**.

4 Select Boundaries 2, 3, 11, and 12 only.



Because of the PMC boundary condition, there is no tangential H-field on the outermost boundaries. It is therefore possible to omit Boundaries 11 and 12 when computing the current even though the model includes those boundaries.

## STUDY 1

### *Step 1: Mode Analysis*

- 1 In the **Model Builder** window, under **Study 1** click **Step 1: Mode Analysis**.
- 2 In the **Settings** window for **Mode Analysis**, locate the **Study Settings** section.
- 3 Select the **Desired number of modes** check box. In the associated text field, type 1.

## DEFINITIONS

### *Variables 1*

- 1 In the **Definitions** toolbar, click  $\partial =$  **Local Variables**.
- 2 In the **Settings** window for **Variables**, locate the **Variables** section.

3 In the table, enter the following settings:

Name	Expression	Unit	Description
V	$\text{int\_E}(-\text{emw.Ex}*\text{t1x}-\text{emw.Ey}*\text{t1y})$	V	Voltage
I	$-\text{int\_H}(\text{emw.Hx}*\text{t1x}+\text{emw.Hy}*\text{t1y})$	A	Current
Z_model	V/I	$\Omega$	Characteristic impedance

Here,  $t1x$  and  $t1y$  are the tangential vector components along the integration boundaries (1 refers to the boundary dimension). The `emw.` prefix gives the correct physics-interface scope for the electric and magnetic field vector components.

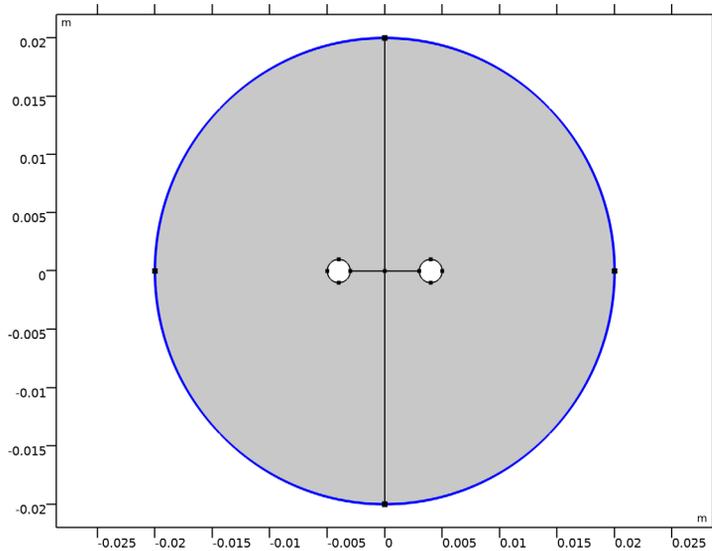
### ELECTROMAGNETIC WAVES, FREQUENCY DOMAIN (EMW)

Now set up the physics. The default boundary condition is perfect electric conductor. Override the outermost boundaries with a perfect magnetic conductor condition to create a virtually infinite modeling space.

#### Perfect Magnetic Conductor 1

1 In the **Model Builder** window, under **Component 1 (comp1)** right-click **Electromagnetic Waves, Frequency Domain (emw)** and choose **Perfect Magnetic Conductor**.

2 Select Boundaries 5, 6, 11, and 12 only.



## MATERIALS

Next, assign a material to the modeling domain.

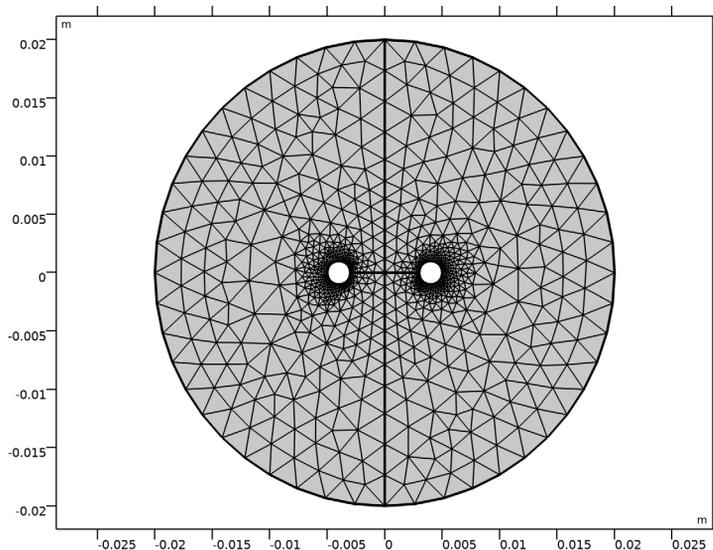
### ADD MATERIAL

- 1 In the **Home** toolbar, click  **Add Material** to open the **Add Material** window.
- 2 Go to the **Add Material** window.
- 3 In the tree, select **Built-in>Air**.
- 4 Click **Add to Component** in the window toolbar.
- 5 In the **Home** toolbar, click  **Add Material** to close the **Add Material** window.

### MESH 1

Use the default mesh.

- 1 In the **Model Builder** window, under **Component 1 (comp1)** click **Mesh 1**.
- 2 In the **Settings** window for **Mesh**, locate the **Physics-Controlled Mesh** section.
- 3 In the table, clear the **Use** check box for **Electromagnetic Waves, Frequency Domain (emw)**.
- 4 Click  **Build All**.



### STUDY 1

In the **Home** toolbar, click  **Compute**.

## RESULTS

### *Electric Field (emw)*

The default plot shows the distribution of the norm of the electric field. Add an arrow plot of the magnetic field.

### *Arrow Surface 1*

- 1 Right-click **Electric Field (emw)** and choose **Arrow Surface**.
- 2 In the **Settings** window for **Arrow Surface**, locate the **Arrow Positioning** section.
- 3 Find the **X grid points** subsection. In the **Points** text field, type 30.
- 4 Find the **Y grid points** subsection. In the **Points** text field, type 30.
- 5 In the **Electric Field (emw)** toolbar, click  **Plot**.
- 6 Locate the **Coloring and Style** section. From the **Color** list, choose **White**.
- 7 Select the **Scale factor** check box. In the associated text field, type  $1.2e-4$ .

You can use the slider to adjust the arrow-length scale factor.

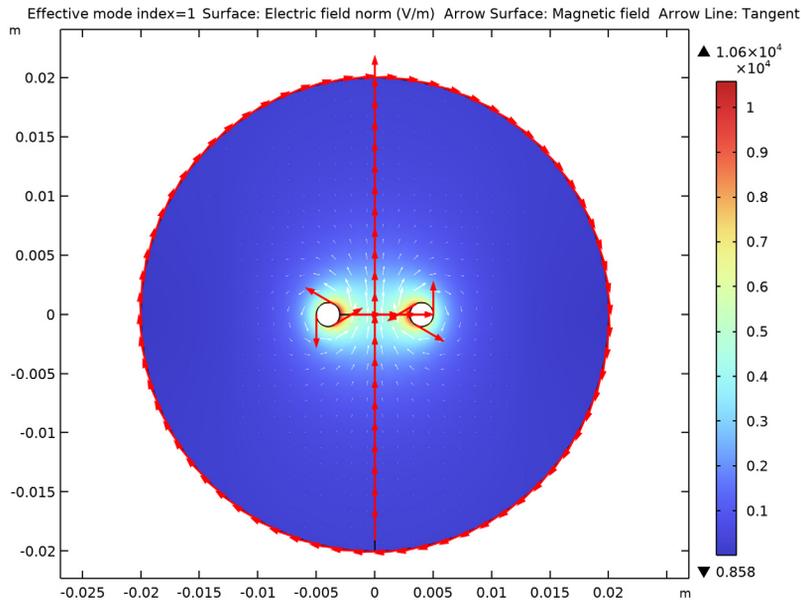
Compare the resulting plot with that shown in [Figure 3](#).

Add an arrow plot along the boundaries to see the orientation of tangent vector field.

### *Arrow Line 1*

- 1 Right-click **Electric Field (emw)** and choose **Arrow Line**.
- 2 In the **Settings** window for **Arrow Line**, click **Replace Expression** in the upper-right corner of the **Expression** section. From the menu, choose **Component 1 (comp 1)>Geometry>tx, ty - Tangent**.
- 3 In the **Electric Field (emw)** toolbar, click  **Plot**.
- 4 Locate the **Arrow Positioning** section. In the **Number of arrows** text field, type 100.

5 In the **Electric Field (emw)** toolbar, click  **Plot**.



A comparison with Equation 1 reveals that the line integral for the voltage computes the potential difference  $V_2 - V_1$ . When computing the line integral for the current, the clockwise orientation of the integration contour would mean that a positive current is directed in the negative  $z$  direction, that is, into the modeling plane. The minus sign added in the definition of  $I$  reverses this direction.

Finish by computing the characteristic impedance.

#### Global Evaluation 1

1 In the **Results** toolbar, click  **Global Evaluation**.

2 In the **Settings** window for **Global Evaluation**, locate the **Expressions** section.

3 In the table, enter the following settings:

Expression	Unit	Description
Z_model	$\Omega$	Characteristic impedance

4 Click  **Evaluate**.

## TABLE

I Go to the **Table** window.

The value shown in the Table window should be close to  $255.8\Omega$ . You can get closer to the analytic value  $247.4\Omega$  by increasing the air-domain radius. For example, changing the definition of `r_air` to `10*r_d` under Global Definitions>Parameters and re-solving the model gives the value  $248.9\Omega$ .

